

EE540 Advance Electromagnetic Theory & Antennas

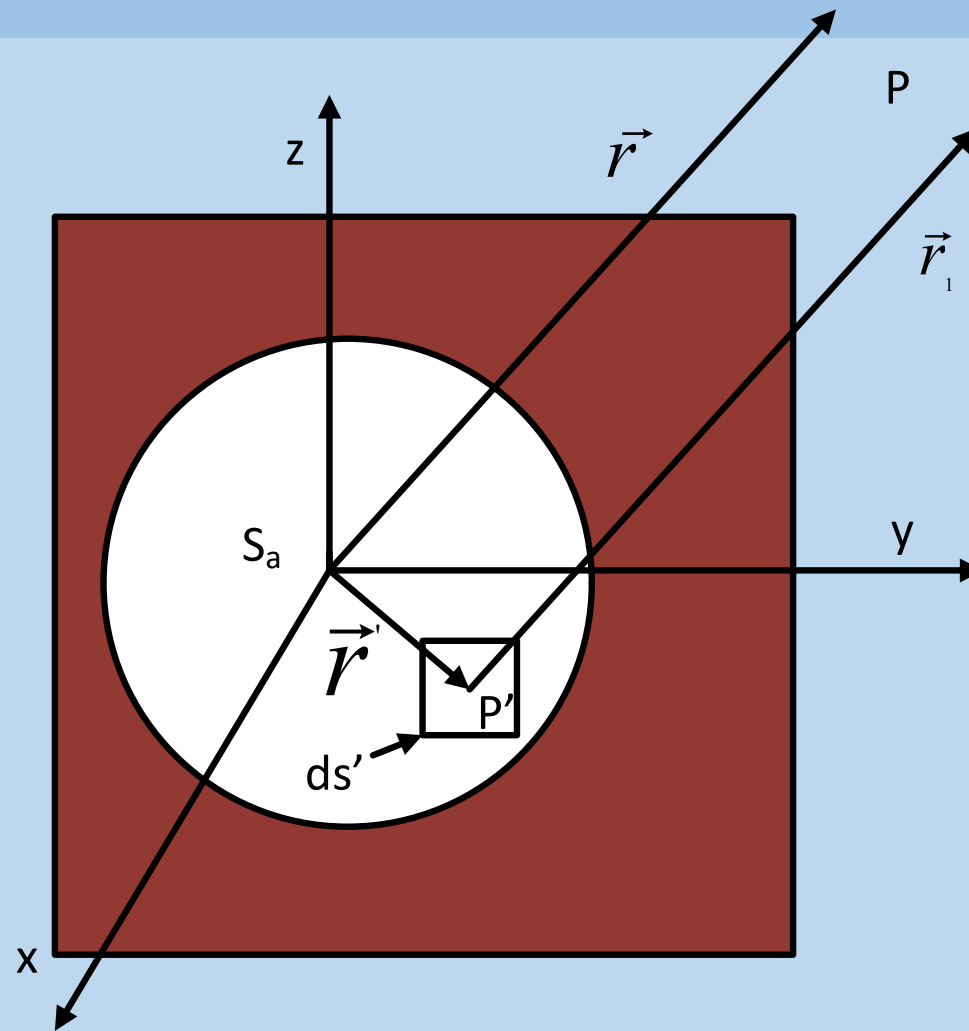
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Aperture antennas

- Two types of antennas:
- Type I (actual source) antennas
 - such as Dipole, Loop antennas
- Type II (equivalent source) antennas
 - such as aperture antennas
- Antenna analysis for Type I antennas:
 - $(\vec{J}_S, \vec{M}_S) \rightarrow (\vec{A}, \vec{F}) \rightarrow (\vec{E}, \vec{H})$
- Antenna analysis for Type II antennas:
 - $(\vec{E}_a, \vec{H}_a) \rightarrow (\vec{J}_S, \vec{M}_S) \rightarrow (\vec{A}, \vec{F}) \rightarrow (\vec{E}, \vec{H})$

Aperture antennas

- Fig. Radiation from Type II antenna (an aperture S_a positioned in x-y plane)



Aperture antennas

- In type II antennas,
 - fields in the aperture (\vec{E}_a, \vec{H}_a) is always known
- For example,
 - in open-ended waveguide,
 - we know the field propagating inside the waveguide
- Hence we can find equivalent currents from the aperture fields as
 - $\vec{J}_S = \hat{n} \times \vec{H}_a, \vec{M}_S = -\hat{n} \times \vec{E}_a$
- So we can find

- $$\vec{A} = \frac{\mu}{4\pi} \iint_{S_a} \vec{J}_S \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} ds'$$

Aperture antennas

- For our case,

- $\vec{J}_S = \hat{n} \times \vec{H}_a = \hat{z} \times (H_{ax}\hat{x} + H_{ay}\hat{y}) = -H_{ay}\hat{x} + H_{ax}\hat{y}$

- So we can find

- $\vec{A} = \frac{\mu}{4\pi} \iint_{S_a} (-H_{ay}\hat{x} + H_{ax}\hat{y}) \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} ds'$

- In far field,

- $|\vec{r} - \vec{r}'| = r$ for amplitude term

- For phase term,

- $|\vec{r} - \vec{r}'| = [(x - x')^2 + (y - y')^2 + (z)^2]^{\frac{1}{2}}$
 $= [r^2 - 2xx' - 2yy' + (x')^2 + (y')^2]^{\frac{1}{2}}$

Aperture antennas

- For phase term,

- $|\vec{r} - \vec{r}'| = [r^2 - 2xx' - 2yy' + (x')^2 + (y')^2]^{\frac{1}{2}}$
 $= r \left[1 - \frac{2xx'}{r^2} - \frac{2yy'}{r^2} + \frac{(x')^2 + (y')^2}{r^2} \right]^{\frac{1}{2}}$

- Doing binomial expansion,

- $|\vec{r} - \vec{r}'| = r - \frac{xx'}{r} - \frac{yy'}{r} + \frac{(x')^2 + (y')^2}{2r} + \dots$

- For far fields, last term can be neglected

$$|\vec{r} - \vec{r}'| \cong r - \frac{xx'}{r} - \frac{yy'}{r}$$

Aperture antennas

- How can we neglect the last term for far fields?
- Note that for far fields, neglecting the last term will give an error in phase approximation as

$$\beta \frac{(x')^2 + (y')^2}{2r_{ff}} \leq \frac{\pi}{8}$$

$$\text{Or, } \frac{(x')^2 + (y')^2}{2r_{ff}} \leq \frac{\lambda}{16}$$

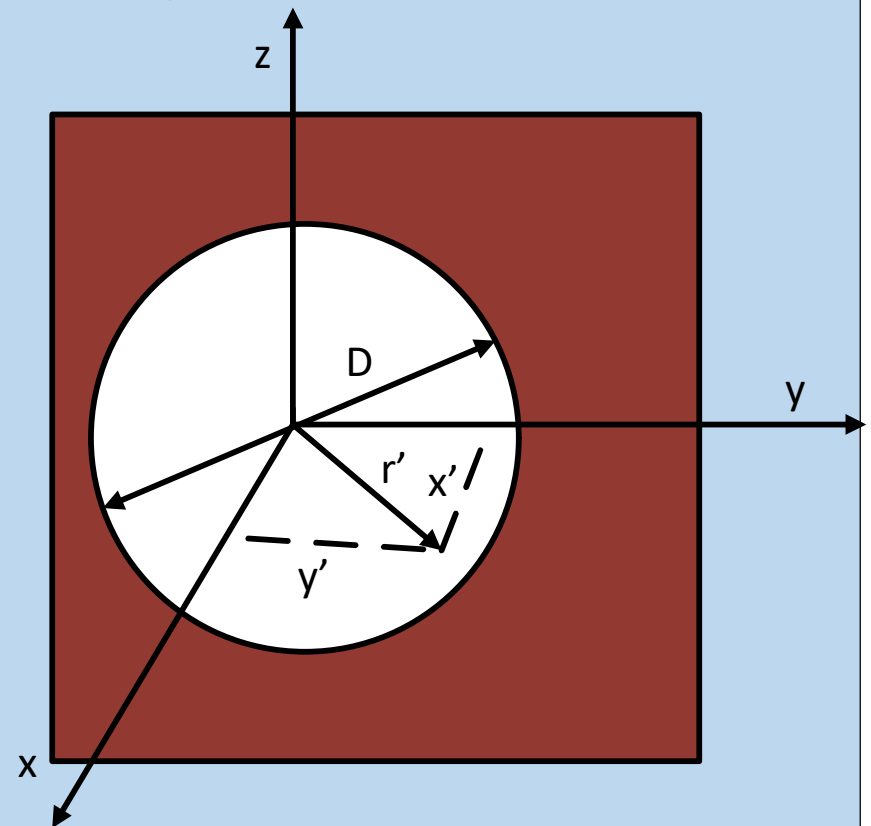
$$\text{Since } r' = \sqrt{(x')^2 + (y')^2} \leq \frac{D}{2}$$

- where D is the largest diameter/dimension of the antenna, therefore,

$$\frac{(D)^2}{8r_{ff}} \leq \frac{\lambda}{16}$$

$$\text{Hence, } \frac{2(D)^2}{\lambda} \leq r_{ff}$$

Fig. Largest dimension of the antenna



Aperture antennas

- Hence,

- $$\frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} \cong \frac{e^{-j\beta\left(r-\frac{xx'}{r}-\frac{yy'}{r}\right)}}{r}$$

- Also note that

- $$\frac{x}{r} = \frac{\beta_x}{\beta} = \sin\theta\cos\phi, \frac{y}{r} = \frac{\beta_y}{\beta} = \sin\theta\sin\phi$$

- Hence

- $$\frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} \cong \frac{e^{(-j\beta r + j\beta_x x' + j\beta_y y')}}{r}$$

Aperture antennas

- Therefore,

- $$\vec{A} = \frac{\mu e^{-j\beta r}}{4\pi r} \iint_{S_a} (-H_{ay}\hat{x} + H_{ax}\hat{y}) e^{(j\beta_x x' + j\beta_y y')} dx' dy'$$

- It can be written as

- $$\vec{A} = \frac{\mu e^{-j\beta r}}{4\pi r} (-\tilde{H}_{ay}\hat{x} + \tilde{H}_{ax}\hat{y})$$

- where \tilde{H}_{ay} and \tilde{H}_{ax} are the 2-D FT of H_{ay} and H_{ax} respectively
- *This is profound and useful result*
 - *which shows that FT plays a significant role in finding fields of type II antenna such as aperture antennas*

Aperture antennas

- Similarly,

- $$\vec{F} = \frac{\epsilon}{4\pi} \iint_{S_a} \vec{M}_S \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} ds'$$

- $$\vec{F} = -\frac{\epsilon e^{(-j\beta r)}}{4\pi r} (-\tilde{E}_{ay}\hat{x} + \tilde{E}_{ax}\hat{y})$$

- where \tilde{E}_{ay} and \tilde{E}_{ax} are the 2-D FT of E_{ay} and E_{ax} respectively

- The negative sign is due to $\vec{M}_S = -\hat{n} \times \vec{E}_a$

- But we always find fields in spherical coordinates

- So we need to do a coordinate transformation

Aperture antennas

- Coordinate transformation from Cartesian \rightarrow Spherical

- $$\begin{pmatrix} A_r/F_r \\ A_\theta/F_\theta \\ A_\phi/F_\phi \end{pmatrix} = \begin{pmatrix} \sin\theta\cos\phi & \sin\theta\sin\phi & \cos\theta \\ \cos\theta\cos\phi & \cos\theta\sin\phi & -\sin\theta \\ -\sin\phi & \cos\phi & 0 \end{pmatrix} \begin{pmatrix} A_x/F_x \\ A_y/F_y \\ A_z/F_z \end{pmatrix}$$

- For far fields, $A_r/F_r \cong 0$

- and for equivalent source in x-y plane, $A_z = F_z = 0$

- $$A_\theta = \frac{\mu e^{-j\beta r}}{4\pi r} (-\tilde{H}_{ay}\cos\theta\cos\phi + \tilde{H}_{ax}\cos\theta\sin\phi)$$

- $$A_\phi = \frac{\mu e^{-j\beta r}}{4\pi r} (\tilde{H}_{ay}\sin\phi + \tilde{H}_{ax}\cos\phi)$$

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- Similarly,
- $F_\theta = -\frac{\epsilon e^{(-j\beta r)}}{4\pi r} (-\tilde{E}_{ay} \cos\theta \cos\phi + \tilde{E}_{ax} \cos\theta \sin\phi)$
- $F_\phi = -\frac{\epsilon e^{(-j\beta r)}}{4\pi r} (\tilde{E}_{ay} \sin\phi + \tilde{E}_{ax} \cos\phi)$
- Electric field in the far field will have contribution from \vec{A} and \vec{F}
- Hence,
- $\vec{E}_{ff} = \vec{E}_{ff}^A + \vec{E}_{ff}^F$
- where $\vec{E}_{ff}^A \cong -j\omega A_\theta \hat{\theta} - j\omega A_\phi \hat{\phi}$
- and $\vec{E}_{ff}^F = \eta \vec{H}_{ff}^F \times \hat{r}$

Aperture antennas

- where $\vec{H}_{ff}^F \cong -j\omega F_\theta \hat{\theta} - j\omega F_\phi \hat{\phi}$
- and $\vec{E}_{ff}^F = -j\omega\eta(F_\theta \hat{\theta} + F_\phi \hat{\phi}) \times \hat{r} = -j\omega\eta(F_\phi \hat{\theta} - F_\theta \hat{\phi})$
- Therefore,
- $E_{ff\theta} \cong -j\omega A_\theta - j\omega\eta F_\phi$
- $E_{ff\phi} \cong -j\omega A_\phi + j\omega\eta F_\theta$
- Hence,
- $$E_{ff\theta} \cong -\frac{j\omega\mu e^{-j\beta r}}{4\pi r} (-\tilde{H}_{ay} \cos\theta \cos\phi + \tilde{H}_{ax} \cos\theta \sin\phi) + \frac{j\omega\eta\epsilon e^{-j\beta r}}{4\pi r} (\tilde{E}_{ay} \sin\phi + \tilde{E}_{ax} \cos\phi)$$

Aperture antennas

- Note that

$$\bullet \quad -\frac{j\omega\mu e^{-j\beta r}}{4\pi r} = -\frac{j\omega\mu\beta e^{-j\beta r}}{\beta 4\pi r} = -\frac{j\eta\beta e^{-j\beta r}}{4\pi r}$$

$$\bullet \quad \frac{j\omega\eta\epsilon e^{-j\beta r}}{4\pi r} = \frac{j\omega\omega\mu\epsilon e^{-j\beta r}}{\beta 4\pi r} = \frac{j\beta^2 e^{-j\beta r}}{\beta 4\pi r} = \frac{j\beta e^{-j\beta r}}{4\pi r}$$

$$\bullet \quad \text{Finally, } E_{ff\theta} \cong \frac{j\beta e^{-j\beta r}}{4\pi r} \left[\eta(\tilde{H}_{ay}\cos\theta\cos\phi - \tilde{H}_{ax}\cos\theta\sin\phi) + (\tilde{E}_{ay}\sin\phi + \tilde{E}_{ax}\cos\phi) \right] \quad \text{---(24.1)}$$

$$\bullet \quad \text{Similarly, } E_{ff\phi} \cong \frac{j\beta e^{-j\beta r}}{4\pi r} \left[-\eta(\tilde{H}_{ay}\sin\phi + \tilde{H}_{ax}\cos\phi) + (\tilde{E}_{ay}\cos\theta\cos\phi - \tilde{E}_{ax}\cos\theta\sin\phi) \right] \quad \text{---(24.2)}$$